

## **General Description**

The MAX8510/MAX8511/MAX8512 ultra-low-noise, lowdropout (LDO) linear regulators are designed to deliver up to 120mA continuous output current from a tiny 5-pin SC70 plastic package. These regulators achieve a low 120mV dropout for 120mA load current. The MAX8510 uses an advanced architecture to achieve ultra-low output voltage noise of  $11\mu$ V<sub>RMS</sub> and PSRR of 54dB at 100kHz.

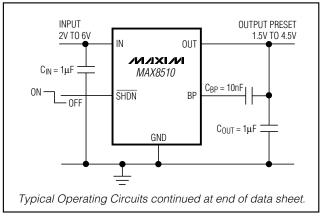
The MAX8511 does not require a bypass capacitor, hence achieving the smallest PC board area. The MAX8512's output voltage can be adjusted with an external divider.

The MAX8510/MAX8511 are preset to a variety of voltages in the 1.5V to 4.5V range. Designed with a P-channel MOSFET series pass transistor, the MAX8510/MAX8511/MAX8512 maintain very low ground current ( $40\mu A$ ).

The regulators are designed and optimized to work with low-value, low-cost ceramic capacitors. The MAX8510 requires only 1 $\mu$ F (typ) of output capacitance for stability with any load. When disabled, current consumption drops to below 1 $\mu$ A.

#### \_Applications

Cellular and Cordless Phones PDA and Palmtop Computers Base Stations Bluetooth Portable Radios and Accessories Wireless LANs Digital Cameras Personal Stereos Portable and Battery-Powered Equipment



# **Typical Operating Circuits**

## 

\_ Maxim Integrated Products 1

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#### **Features**

- Space-Saving 5-Pin SC70 Package
- ♦ 11µV<sub>RMS</sub> Output Noise at 100Hz to 100kHz Bandwidth (MAX8510)
- 78dB PSRR at 1kHz (MAX8510)
- 120mV Dropout at 120mA Load
- ♦ Stable with 1µF Ceramic Capacitor for Any Load
- Guaranteed 120mA Output
- Only Need Input and Output Capacitors (MAX8511)
- Output Voltages: 1.5V, 1.8V, 2.5V, 2.6V, 2.7V, 2.8V, 2.85V, 3V, 3.3V, 4.5V (MAX8510/MAX8511) and Adjustable (MAX8512)
- Low 40µA Ground Current
- Excellent Load/Line Transient
- Overcurrent and Thermal Protection

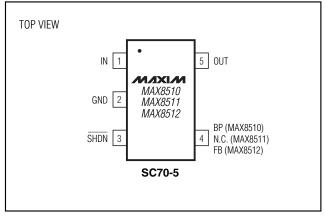
## Ordering Information

TEMP RANGE	PIN-PACKAGE
-40°C to +85°C	5 SC70-5
-40°C to +85°C	5 SC70-5
-40°C to +85°C	5 SC70-5
	-40°C to +85°C -40°C to +85°C

\*xy is the output voltage code (see Output Voltage Selector Guide). Other versions between 1.5V and 4.5V are available in 100mV increments. Contact factory for other versions.

Output Voltage Selector Guide appears at end of data sheet.

# Pin Configuration



## **ABSOLUTE MAXIMUM RATINGS**

IN to GND	0.3V to +7V
Output Short-Circuit Duration	Infinite
OUT, SHDN to GND	0.3V to (IN + 0.3V)
FB, BP, N.C. to GND	
Continuous Power Dissipation ( $T_A = +70$	O°C)
5-Pin SC70 (derate 3.1mW/°C above	+70°C)0.247W

θυα	
(5-Pin SC70)	324°C/W
Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

# **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = V_{OUT} + 0.5V, T_A = -40^{\circ}C$  to +85°C, unless otherwise noted.  $C_{IN} = 1\mu$ F,  $C_{OUT} = 1\mu$ F,  $C_{BP} = 10$ nF. Typical values are at +25°C; the MAX8512 is tested with 2.45V output, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS
Input Voltage Range	VIN					6	V
		$I_{OUT} = 1$ mA, $T_A = +25$ °C	$I_{OUT} = 1$ mA, $T_A = +25$ °C			+1	%
Output Voltage Accuracy		$I_{OUT} = 100\mu A$ to 80mA, $T_A = +25^{\circ}C$		-2		+2	
		$I_{OUT} = 100 \mu A \text{ to } 80 \text{mA}$				+3	
Maximum Output Current	Iout			120			mA
Current Limit	ILIM	OUT = 90% of nominal val	ue	130	200	300	mA
		V <sub>OUT</sub> ≥ 3V, I <sub>OUT</sub> = 80mA			80	170	
		V <sub>OUT</sub> ≥ 3V, I <sub>OUT</sub> = 120mA			120		
		$2.5V \le V_{OUT} < 3V, I_{OUT} = 8$	80mA		90	200	
Dropout Voltage (Note 2)		$2.5V \le V_{OUT} < 3V$ , $I_{OUT} = 120mA$ $2V \le V_{OUT} < 2.5V$ , $I_{OUT} = 80mA$ $2V \le V_{OUT} < 2.5V$ , $I_{OUT} = 120mA$			135		mV
					120	250	
					180		
Ourse of Ourse at		I <sub>OUT</sub> = 0.05mA			40	90	
Ground Current	lq	$V_{IN} = V_{OUT} (nom) - 0.1V, I_{OUT} (nom)$	DUT = 0mA	220		500	μΑ
Line Regulation	V <sub>LNR</sub>	$V_{IN} = (V_{OUT} + 0.5V)$ to 6V, $I_{OUT} = 0.1mA$			0.001		%/V
Load Regulation	VLDR	I <sub>OUT</sub> = 1mA to 80mA			0.003		%/mA
			$T_A = +25^{\circ}C$		0.003	1	μΑ
Shutdown Supply Current	ISHDN	SHDN = 0V	$T_A = +85^{\circ}C$		0.05		
Ripple Rejection			MAX8510		78		dB
		$f = 1$ kHz, $I_{OUT} = 10$ mA	MAX8511/MAX8512		72		
	PSRR	f = 10kHz, I <sub>OUT</sub> = 10mA	MAX8510		75		
			MAX8511/MAX8512		65		
			MAX8510	İ	54		
		$f = 100$ kHz, $I_{OUT} = 10$ mA	MAX8511/ MAX8512		46		

# **ELECTRICAL CHARACTERISTICS (continued)**

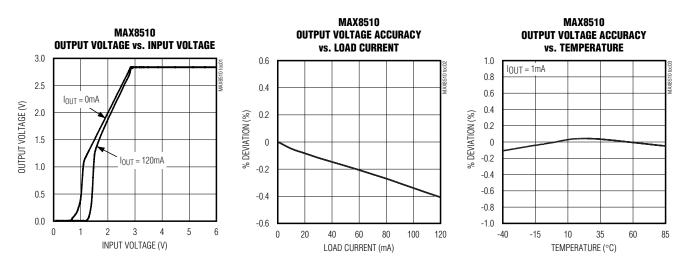
 $(V_{IN} = V_{OUT} + 0.5V, T_A = -40^{\circ}C$  to +85°C, unless otherwise noted.  $C_{IN} = 1\mu$ F,  $C_{OUT} = 1\mu$ F,  $C_{BP} = 10$ nF. Typical values are at +25°C; the MAX8512 is tested with 2.45V output, unless otherwise noted.) (Note 1)

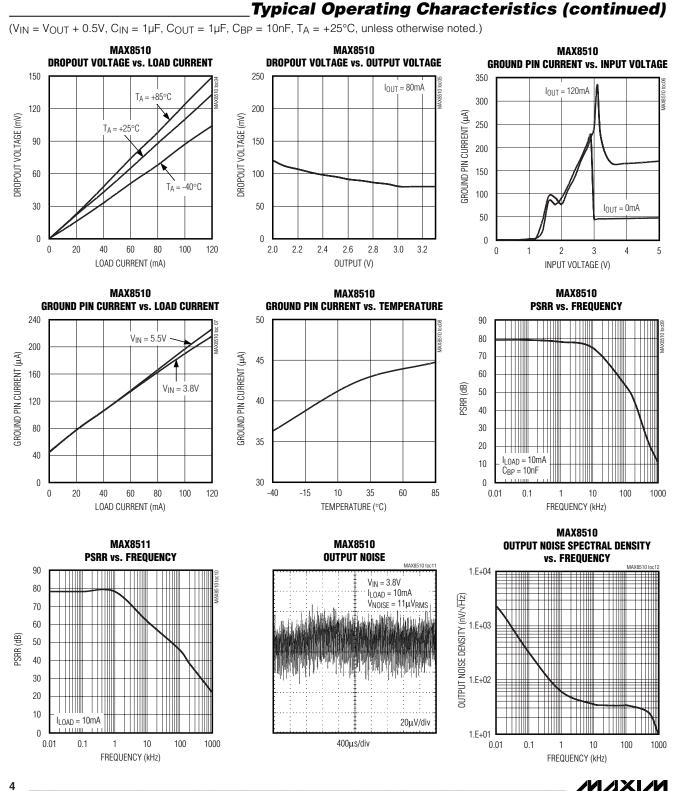
PARAMETER	SYMBOL	CONDITIONS			ТҮР	MAX	UNITS	
Output Noise Voltage		f = 100Hz to 100kHz, I <sub>LOAD</sub> = 10mA	MAX8510		11			
			MAX8511/MAX8512		230		μV	
(RMS)		f = 100Hz to 100kHz,	MAX8510		13			
		$I_{LOAD} = 80 mA$	MAX8511/MAX8512		230			
Shutdown Exit Delay		$R_{LOAD} = 50\Omega$ (Note3)			300	μs		
SHDN Logic Low Level		$V_{IN} = 2V$ to $6V$			0.4	V		
SHDN Logic High Level		$V_{IN} = 2V$ to $6V$	1.5	0.001	0.1	V		
		$V_{IN} = 6V, \overline{SHDN} = 0V \text{ or}$	$T_A = +25^{\circ}C$				μA	
SHDN Input Bias Current		6V	$T_A = +85^{\circ}C$		0.01			
B Input Bias Current			$T_A = +25^{\circ}C$		0.006	0.1		
(MAX8512)		$V_{\rm IN} = 6V, V_{\rm FB} = 1.3V$	$T_A = +85^{\circ}C$		0.01	μΑ		
Thermal Shutdown					160		°C	
Thermal-Shutdown Hysteresis					10		°C	

**Note 1:** Limits are 100% tested at +25°C. Limits over operating temperature range are guaranteed by design. **Note 2:** Dropout is defined as  $V_{IN} - V_{OUT}$  when  $V_{OUT}$  is 100mV below the value of  $V_{OUT}$  for  $V_{IN} = V_{OUT} + 0.5V$ . **Note 3:** Time needed for  $V_{OUT}$  to reach 90% of final value.

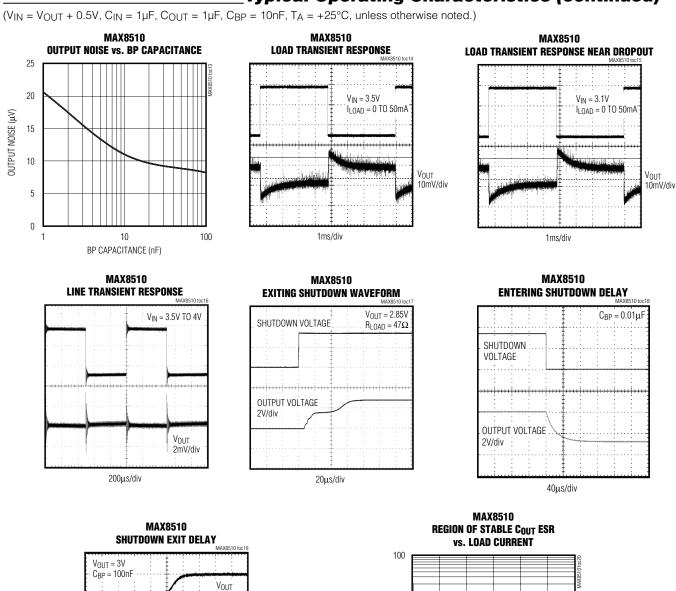
# **Typical Operating Characteristics**

(VIN = VOUT + 0.5V, CIN = 1 $\mu$ F, COUT = 1 $\mu$ F, CBP = 10nF, TA = +25°C, unless otherwise noted.)





MAX8510/MAX8511/MAX8512



10

1

0.1

0.01

0 20

STABLE REGION

LOAD CURRENT (mA)

80 100

120

40 60

 $C_{OUT}$  ESR ( $\Omega$ )

1V/div

...........

SHUTDOWN VOLTAGE

20µs/div

## Typical Operating Characteristics (continued)

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MAX8510/MAX8511/MAX8512

## Pin Description

PIN			NAME	FUNCTION	
MAX8510	MAX8511	MAX8512	NAME	FUNCTION	
1	1	1	IN	Unregulated Input Supply	
2	2	2	GND	Ground	
3	3	3	SHDN	Shutdown. Pull low to disable the regulator.	
4	_	_	BP Noise Bypass for Low-Noise Operation. Connect a 10nF capacitor fro BP to OUT. It is short to OUT in shutdown mode.		
_	4	_	N.C. Not Internally Connected		
_	_	4	FB Adjustable Output Feedback Point		
5	5	5	OUT Regulated Output Voltage. Bypass with a capacitor to GND. See the <i>Capacitor Selection and Regulator Stability</i> section for more details.		

## **Detailed Description**

The MAX8510/MAX8511/MAX8512 are ultra-low-noise, low-dropout, low-quiescent current linear regulators designed for space-restricted applications. The parts are available with preset output voltages ranging from 1.5V to 4.5V in 100mV increments. These devices can supply loads up to 120mA. As shown in the *Functional Diagram*, the MAX8510/MAX8511 consist of an innovative bandgap core and noise bypass circuit, error amplifier, P-channel pass transistor, and internal feedback voltage-divider. The MAX8512 allows for adjustable output with an external feedback network.

The 1.225V bandgap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the feedback voltage and amplifies the difference. If the feedback voltage is lower than the reference voltage, the pass-transistor gate is pulled low. This allows more current to pass to the output and increases the output voltage. If the feedback voltage is pulled high, allowing less current to pass to the output. The output voltage is fed back through an internal resistor voltage-divider connected to the OUT pin.

An external bypass capacitor connected to BP (MAX8510) reduces noise at the output. Additional blocks include a current limiter, thermal sensor, and shutdown logic.

#### Internal P-Channel Pass Transistor

The MAX8510/MAX8511/MAX8512 feature a 1 $\Omega$  (typ) Pchannel MOSFET pass transistor. This provides several advantages over similar designs using a PNP pass transistor, including longer battery life. The P-channel MOSFET requires no base drive, which considerably reduces quiescent current. PNP-based regulators waste considerable current in dropout when the pass transistor saturates. They also use high base-drive current under heavy loads. The MAX8510/MAX8511/MAX8512 do not suffer from these problems and consume only  $40\mu$ A of quiescent current in light load and  $220\mu$ A in dropout (see the *Typical Operating Characteristics*).

#### **Output Voltage Selection**

The MAX8510/MAX8511 are supplied with factory-set output voltages from 1.5V to 4.5V, in 100mV increments (see *Ordering Information*). The MAX8512 features a user-adjustable output through an external feedback network (see the *Typical Operating Circuits*).

To set the output of the MAX8512, use the following equation:

$$R1 = R2 X \left(\frac{V_{OUT}}{V_{REF}} - 1\right)$$

where R2 is chosen to be less than 240k $\Omega$  and V<sub>REF</sub> = 1.225V. Use 1% or better resistors.

#### Shutdown

The MAX8510/MAX8511/MAX8512 feature a low-power shutdown mode that reduces quiescent current less than 1 $\mu$ A. Driving SHDN low disables the voltage reference, error amplifier, gate-drive circuitry, and pass transistor (see the *Functional Diagram*), and the device output enters a high-impedance state. Connect SHDN to IN for normal operation.

#### **Current Limit**

The MAX8510/MAX8511/MAX8512 include a current limiter, which monitors and controls the pass transistor's gate voltage, limiting the output current to 200mA.



For design purposes, consider the current limit to be 130mA (min) to 300mA (max). The output can be shorted to ground for an indefinite amount of time without damaging the part.

#### **Thermal-Overload Protection**

Thermal-overload protection limits total power dissipation in the MAX8510/MAX8511/MAX8512. When the junction temperature exceeds  $T_J = +160^{\circ}$ C, the thermal sensor signals the shutdown logic, turning off the pass transistor and allowing the IC to cool down. The thermal sensor turns the pass transistor on again after the IC's junction temperature drops by 10°C, resulting in a pulsed output during continuous thermal-overload conditions.

Thermal-overload protection is designed to protect the MAX8510/MAX8511/MAX8512 in the event of a fault condition. For continual operation, do not exceed the absolute maximum junction temperature rating of  $T_J$  = +150°C.

**Operating Region and Power Dissipation** 

The MAX8510/MAX8511/MAX8512 maximum power dissipation depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient, and the rate of airflow. The power dissipation across the device is:

P = IOUT (VIN - VOUT)

The maximum power dissipation is:

 $P_{MAX} = (T_J - T_A) / (\theta_{JC} + \theta_{CA})$ 

where T<sub>J</sub> - T<sub>A</sub> is the temperature difference between the MAX8510/MAX8511/MAX8512 die junction and the surrounding air,  $\theta_{JC}$  is the thermal resistance of the package, and  $\theta_{CA}$  is the thermal resistance through the PC board, copper traces, and other materials to the surrounding air.

The GND pin of the MAX8510/MAX8511/MAX8512 performs the dual function of providing an electrical connection to ground and channeling heat away. Connect the GND pin to ground using a large pad or ground plane.

#### **Noise Reduction**

For the MAX8510, an external  $0.01\mu$ F bypass capacitor between BP and OUT with innovative noise bypass scheme reduces output noises dramatically, exhibiting  $11\mu$ V<sub>RMS</sub> of output voltage noise with C<sub>BP</sub> =  $0.01\mu$ F and C<sub>OUT</sub> =  $1\mu$ F. Startup time is minimized by a poweron circuit that precharges the bypass capacitor.

#### \_Applications Information

#### Capacitor Selection and Regulator Stability

Use a 1 $\mu$ F capacitor on the MAX8510/MAX8511/ MAX8512 input and a 1 $\mu$ F capacitor on the output. Larger input capacitor values and lower ESRs provide better noise rejection and line-transient response. Reduce output noise and improve load-transient response, stability, and power-supply rejection by using large output capacitors. Note that some ceramic dielectrics exhibit large capacitance and ESR variation with temperature. With dielectrics such as Z5U and Y5V, it may be necessary to use a 2.2 $\mu$ F or larger output capacitor to ensure stability at temperatures below -10°C. With X7R or X5R dielectrics, 1 $\mu$ F is sufficient at all operating temperatures. A graph of the region of stable COUT ESR vs. load current is shown in the *Typical Operating Characteristics*.

Use a 0.01 $\mu$ F bypass capacitor at BP (MAX8510) for low-output voltage noise. The leakage current going into the BP pin should be less than 10nA. Increasing the capacitance slightly decreases the output noise. Values above 0.1 $\mu$ F and below 0.001 $\mu$ F are not recommended.

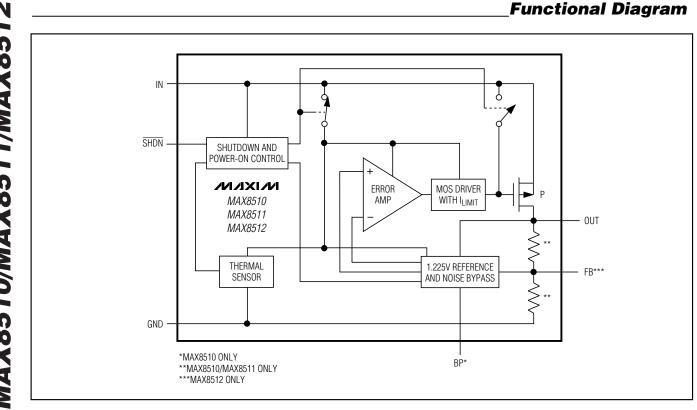
#### Noise, PSRR, and Transient Response

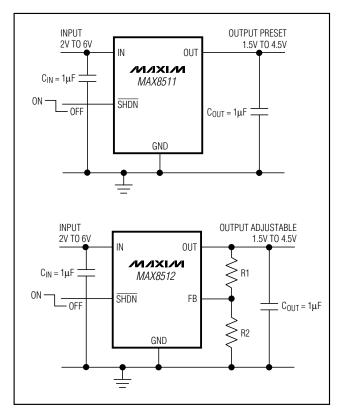
The MAX8510/MAX8511/MAX8512 are designed to deliver ultra-low noise and high PSRR, as well as low dropout and low quiescent currents in battery-powered systems. The MAX8510 power-supply rejection is 78dB at 1kHz and 54dB at 100kHz. The MAX8511/MAX8512 PSRR is 72dB at 1kHz and 46dB at 100kHz (see the Power-Supply Rejection Ratio vs. Frequency graph in the *Typical Operating Characteristics*).

When operating from sources other than batteries, improved supply-noise rejection and transient response can be achieved by increasing the values of the input and output bypass capacitors, and through passive filtering techniques. The *Typical Operating Characteristics* show the MAX8510/MAX8511/MAX8512 line- and loadtransient responses.

#### **Dropout Voltage**

A regulator's minimum dropout voltage determines the lowest usable supply voltage. In battery-powered systems, this determines the useful end-of-life battery voltage. Because the MAX8510/MAX8511/MAX8512 use a P-channel MOSFET pass transistor, their dropout voltage is a function of drain-to-source on-resistance (RDS(ON)) multiplied by the load current (see the *Typical Operating Characteristics*).





## Typical Operating Circuits (continued)

# Output Voltage Selector Guide

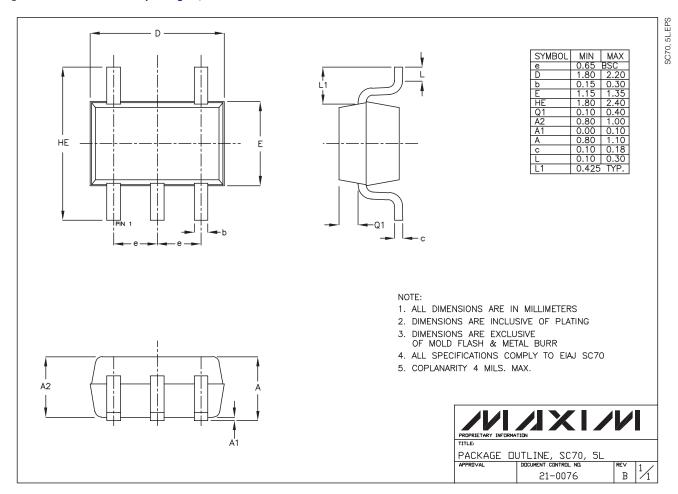
PART	Vout (V)	TOP MARK
MAX8510EXK18-T	1.8	AEA
MAX8510EXK25-T	2.5	AEB
MAX8510EXK27-T	2.7	AEC
MAX8510EXK28-T	2.8	AED
MAX8510EXK29-T	2.85	ADS
MAX8510EXK30-T	3	ADT
MAX8510EXK33-T	3.3	AEE
MAX8511EXK15-T	1.5	ADU
MAX8511EXK18-T	1.8	AEF
MAX8511EXK25-T	2.5	ADV
MAX8511EXK26-T	2.6	AEG
MAX8511EXK29-T	2.85	AEH
MAX8511EXK33-T	3.3	AEI
MAX8511EXK45-T	4.5	AEJ
MAX8512EXK-T	Adjustable	ADW

### **Chip Information**

TRANSISTOR COUNT: 284 PROCESS: BICMOS

## **Package Information**

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to <u>www.maxim-ic.com/packages</u>.)



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